

The Pulse Designing in UWB System and a Novel Wavelet/Sinuquotient

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Introduction

In the reference [1], a generating-TDMA-signal method-Carrier Interferometry-CI is proposed. In using the Dobecci basic wavelet to detecting ultra-wideband signals in noise, each wavelet can be a UWB signal by definition [2]. There are many papers, which deal with UWB signals with wavelets such as $g(t) = e^{-jt} e^{-t^2}$ [3]. This paper gives a UWB signal generating method with the extended trigonometric function wavelet -sinuquotient. That CI method can generate a UWB signal gives light to the author to study the related pulse shape. Different from the conventional trigonometric function wavelet, the paper gives a $\sin(\omega x)/x$ and $(1-\cos\omega x)/x$ as a novel wavelet-sinuquotient. The paper analyzes the characteristics of the sinuquotient, and discusses it as a new tool to deal with the UWB signal.

The characteristics of sinuquotient

The CI method (generating TDMA signal): $\sum_{i=1}^N A \cos(i2\pi\Delta f t)$, $1 \leq i \leq N$ and

obtaining a signal as following:

$$\sum_{i=1}^N A \cos(i2\pi\Delta f t) = A \times \frac{\sin(\frac{N}{2} 2\pi\Delta f t)}{\sin(\frac{1}{2} 2\pi\Delta f t)} \times \cos(\frac{N+1}{2} 2\pi\Delta f t) = A \times \frac{\sin((N+\frac{1}{2}) \times 2\pi\Delta f t) - \sin(\frac{1}{2} \times 2\pi\Delta f t)}{2\sin(\frac{1}{2} \times 2\pi\Delta f t)} \quad (1)$$

Where $A = \sqrt{1/N} \sqrt{2/T_s}$.

In designing the optimum UWB signal, we adopt a method like CI method to generate the UWB signal, which is according to FCC mask. The method is different from the former method that the parameter A is 1. It is not only simplifying processing but also giving convenience in synthesizing signals.

It is supposed that the minimum-frequency triangle function has the $\Delta f/2$ frequency. The upper frequency and the lower frequency must be integer times of $\Delta f/2$; otherwise it will generate ∞ in the pulse. The result is obtained as following (mark $N \times 2\pi \times \Delta f = \omega$):

$$\sum_{i=1}^N \cos(i \times 2\pi \times \Delta f \times t) = \frac{\sin((N+\frac{1}{2}) \times 2\pi\Delta f t) - \sin(\frac{1}{2} \times 2\pi\Delta f t)}{2\sin(\frac{1}{2} \times 2\pi\Delta f t)} = \frac{\sin(\omega + \frac{1}{2} \times 2\pi\Delta f t) - \sin(\frac{1}{2} \times 2\pi\Delta f t)}{2\sin(\frac{1}{2} \times 2\pi\Delta f t)} \quad (2a),$$

$$\sum_{i=1}^N \sin(i \times 2\pi \times \Delta f \times t) = \frac{\cos(\frac{1}{2} \times 2\pi\Delta f t) - \cos((N+\frac{1}{2}) \times 2\pi\Delta f t)}{2\sin(\frac{1}{2} \times 2\pi\Delta f t)} = \frac{\cos(\frac{1}{2} \times 2\pi\Delta f t) - \cos(\omega + \frac{1}{2} \times 2\pi\Delta f t)}{2\sin(\frac{1}{2} \times 2\pi\Delta f t)} \quad (2b)$$

When Δf reducing, the pulse is becoming sparse, and it can be considered as only 1 pulse when $\Delta f \rightarrow 0$. We have the following formula approximately (K and P are constant): $\lim_{\Delta f \rightarrow 0} \sum_{i=1}^N \cos(i \times 2\pi \times \Delta f \times t) \rightarrow K \frac{\sin \omega t}{t}$, $\lim_{\Delta f \rightarrow 0} \sum_{i=1}^N \sin(i \times 2\pi \times \Delta f \times t) \rightarrow P \frac{1 - \cos \omega t}{t}$ (3)

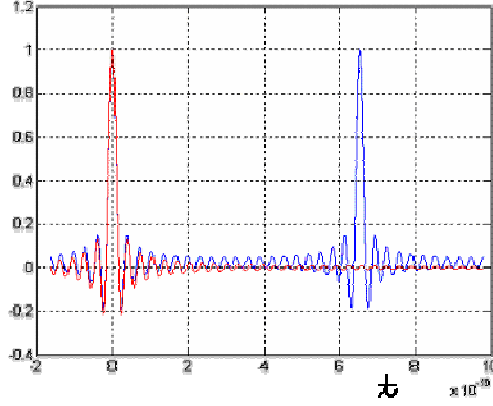


Figure 1 $\sum_{i=1}^N \cos(i \times 2\pi \times \Delta f \times t)$ and $\frac{\sin \omega t}{t}$

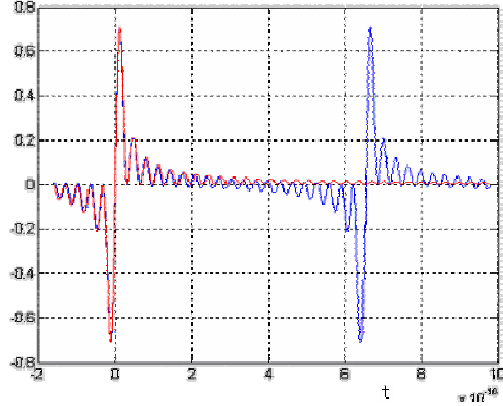


Figure 2 $\sum_{i=1}^N \sin(i \times 2\pi \times \Delta f \times t)$ and $\frac{1 - \cos \omega t}{t}$

Thus analogy is elicited: $\cos \omega t \leftrightarrow \frac{\sin \omega t}{t}$ (4a); $\sin \omega t \leftrightarrow \frac{1 - \cos \omega t}{t}$ (4b). Thus:

$$\cos[\omega(t+\tau) - \theta] \leftrightarrow \frac{\sin \omega(t-\tau)}{t} \cos \theta + \frac{1 - \cos \omega(t-\tau)}{t} \sin \theta = \frac{\sin[\omega(t+\tau) - \theta] + \sin \theta}{t + \tau} \quad (5)$$

The author defines $\frac{\sin[\omega(t+\tau) - \theta] + \sin \theta}{t + \tau}$ sinuquotient, one extended trigonometric function wavelet, and $\frac{\sin \omega t}{t}$ and $\frac{1 - \cos \omega t}{t}$ are defined sine sinuquotient and cosine sinuquotient respectively. The conventional trigonometric function wavelet has only one mother wavelet $\frac{\sin \omega t}{\pi t}$.

The following is the analysis of sine sinuquotient and cosine sinuquotient
1 when $\omega=0$, $\frac{\sin \omega t}{t} \equiv 0$, $\frac{1 - \cos \omega t}{t} \equiv 0$, so $\omega \neq 0$.

2 in the frequency domain, Fourier transform of $\frac{\sin \omega t}{t}$, $F(\frac{\sin \omega t}{t})$ is a gate function. We define a gate function as the following form:

$$G(x_1, x_2) = 1, \forall x \in [x_1, x_2], G(x_1, x_2) = 0 \text{ else } x \quad (6)$$

$$F(\frac{\sin \omega t}{t}) = \pi G(-\omega, \omega), F(\frac{1 - \cos \omega t}{t}) = j\pi[G(0, \omega) - G(-\omega, 0)], \omega \neq 0$$

3 sinuquotient is belong to L^2 , $\int_{-\infty}^{\infty} [\frac{\sin[\omega(t+\tau) - \theta] + \sin \theta}{t + \tau}]^2 dt < \infty$

4 sine sinuquotient and cosine sinuquotient have orthogonality.

$$\int_{-\infty}^{\infty} \frac{\sin \omega_1 t}{t} \times \frac{\sin \omega_2 t}{t} dt = \int_{-\infty}^{\infty} \frac{1}{2\pi} \int_{-\infty}^{\infty} \pi G(-\omega_1, \omega_1) e^{j\omega t} d\omega \times \frac{\sin \omega_2 t}{t} dt = \pi \min(\omega_1, \omega_2) \quad (7)$$

From the above, $\frac{\sin \omega_1 t - \sin \omega_2 t}{t}$ is orthogonal with the $\frac{\sin \omega_1 t}{t}$ ($\omega_1 > \omega_2$). And it is also proven that $A=1$ is reasonable. And cosine sinuquotient as well as sine sinuquotient has the same orthogonality.

5 sinuquotient and its translation

$$\int_{-\infty}^{\infty} \frac{\sin \omega_1 t}{t} \times \frac{\sin \omega_0(t+\tau)}{t+\tau} dt = \frac{\pi \sin \omega_0 \tau}{\tau}, \int_{-\infty}^{\infty} \frac{1 - \cos \omega_0 t}{t} \times \frac{1 - \cos \omega_0(t+\tau)}{t+\tau} dt = \frac{\pi \sin \omega_0 \tau}{\tau}$$

$$\int_{-\infty}^{\infty} \frac{1 - \cos \omega_1 t}{t} \times \frac{\sin \omega_0(t+\tau)}{t+\tau} dt = -\frac{\pi(1 - \cos \omega_0 \tau)}{\tau} \quad (8)$$

Let $\tau=0$, the energy of sinuquotient is $\pi\omega$.

6 the relationship between sinuquotient and sinusoid.

Choose another function $\frac{\sin(\omega - \Delta)t}{t}$ to be subtracted from $\frac{\sin \omega t}{t}$:

$$\frac{\sin \omega t}{t} - \frac{\sin(\omega - \Delta)t}{t} = \frac{\sin \omega t - \sin(\omega - \Delta)t}{t} = \frac{2 \cos(\omega - \frac{1}{2}\Delta)t \sin \frac{1}{2}\Delta \times t}{t} \quad (9)$$

when $\Delta \rightarrow 0$, $\frac{\sin \omega t}{t} - \frac{\sin(\omega - \Delta)t}{t} \cong \Delta \sin \omega t \left(\frac{1 - \cos \omega t}{t} - \frac{1 - \cos(\omega - \Delta)t}{t} \right) \cong \Delta \sin \omega t$.

7 sinuquotient vs sampling function $\delta(t)$ and Hilbert transform.

Choose a function $f(t)$ and its $F(j\omega)$ has upper frequency f_h , and suppose $F_{\omega_0}(j\omega)$ is the component of $F(j\omega)$ in $\omega \leq \omega_0$ and $F_{\omega_0}^{-1}(j\omega) = f_{\omega_0}(t)$. Hilbert transform of $f_{\omega_0}(t)$ is $f_{\omega_0}^h(t)$, $F_{\omega_0}^h(j\omega)$ is the Fourier transform of $f_{\omega_0}^h(t)$:

$$\int_{-\infty}^{\infty} f(t) \frac{\sin(\omega_0 t)}{t} dt = \pi f_{\omega_0}(0) \quad (12a), \int_{-\infty}^{\infty} f(t) \frac{1 - \cos \omega t}{t} dt = \pi f_{\omega_0}^h(0) \quad (10)$$

8 the energy feathers of sinuquotient

The table 1 gives the energy of sinuquotient in limited time (the number of sampling in time is 10000 in $[0, 2\pi/\omega]$). From the table, the sinuquotient can be truncated in $[-5\pi/\omega, 5\pi/\omega]$ with little error.

\Energy (100%)t	$[-\pi/\omega, \pi/\omega]$	$[-2\pi/\omega, 2\pi/\omega]$	$[-3\pi/\omega, 3\pi/\omega]$	$[-4\pi/\omega, 4\pi/\omega]$	$[-5\pi/\omega, 5\pi/\omega]$
$\int \left[\frac{\sin \omega t}{t} \right]^2 dt$	90.282%	94.994%	96.641%	97.475%	97.978%
$\int \left[\frac{1 - \cos \omega t}{t} \right]^2 dt$	65.262%	85.571%	89.578%	92.513%	93.854%

Table 1 the energy feathers of sinuquotient

Sinuquotient in UWB signal processing

Given there are 9 UWB signals (3 superposition), which have the same bandwidth 500MHz. There are 3 lines, center frequency 3.3GHz(Δ tag), 3 lines, center frequency 4.0GHz($\square$ tag), and 3 lines, center frequency 4.55GHz($\square$ tag). The synthesized signal is plotted in the upper of figure 3 and 4. According to the sinuquotient orthogonality, if we choose a sinuquotient signal as the signal tagged by Δ to multiply the synthesized signal and integrate the product, the bandwidth-

500MHz-center frequency-3.3GHz component will be collected and others will be cut. The simulated result is plotted in figure 3 and 4.

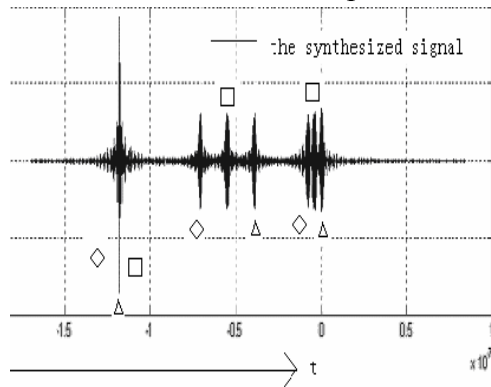


figure 3 done by sine sinuquotient

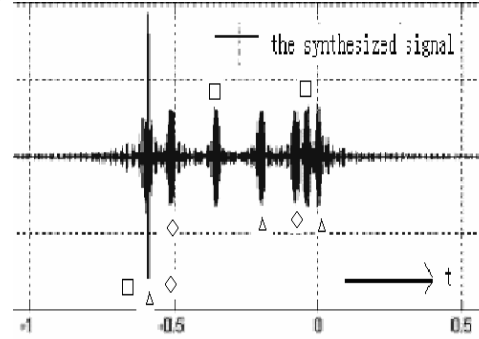


figure 4 done by cosine sinuquotient

Conclusion

The sinuquotient is a novel wavelet and it is simpler than other wavelets in frequency domain and has many perfect traits. And sinuquotient is a convenient tool to process the UWB signals. How to apply the sinuquotient to the practice is the further work in the future.

References:

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- [2] K.Pourvoyeur, A.Stelzer "Wavelet-based impulse reconstruction in UWB-radar", 2003 IEEE MTT-S Digest
- [3] L. F. Chernogor and O.V. Lazorenko "Application of the wavelet analysis for detecting ultra-wideband signals in noise", MMET 2000 Proceedings